

# ***U.S. PATENT APPLICATION***

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***Invention:*** KNOCKING DETECTING DEVICE FOR AUTOMOBILE

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# **KNOCKING DETECTING DEVICE FOR AUTOMOBILE**

## **CROSS-REFERENCE TO RELATED APPLICATION**

This application is based on Japanese Patent Application No.  
5 2003-102027 filed April 04, 2003, the disclosure of which is  
incorporated herein by reference.

## **FIELD OF THE INVENTION**

The present invention relates to a knocking detecting device  
10 for an automobile.

## **BACKGROUND OF THE INVENTION**

Japanese patent No. 2764495 shows a conventional knocking  
detecting device for an automobile. In this device, an analog signal  
15 output from a knock sensor is analog - to - digital converted in  
a constant sampling period and the converted data (sampling data)  
on the time series are digital filter processed. Then, a knocking  
is detected base on the sampling data. An analog - to - digital  
conversion is referred as A/D conversion and an analog - to - digital  
20 converter is referred as A/D converter herein after.

A knocking detection is processed according to all  
filter-processed data which are saved every sampling timing during  
a knocking detection period. During knocking detection, a knock  
sensor signal is processed and the crankshaft of an internal  
25 combustion engine rotates in a predetermined angle.

Since the device takes a long time to detect the knocking in  
low rotation of an engine, a large number of the data should be

stored so that a memory capacity and processing load are increased.

To avoid such a problem, it is suggested that a filter-processed data are integrated to every fixed number  $N$  and the existence of the knocking is detected from the integrated value to reduce the stored data. In such a process, the integrated filter processed data can be deleted and only integrated data are stored so that which the number of the stored data can be reduced to " $1 / K$ ". As shown in Fig. 13, the knocking detection is carried out according to the fluctuation of the integrated value.

10        The device described above needs a long period to detect knocking and the number of stored data (the number of integrated number) is increased in a low rotation of the engine. If the value  $N$  is increased to decrease the number of integrated value, the number of integrated value is reduced in a high rotation of the engine and the knocking detection cannot be carried out precisely. That is, if the number of the integrated value is too small, the characteristic of the knocking cannot be detected.

20        To avoid such a problem, it suggested that A/D converted and digital-filtered data are integrated every period in which a crankshaft rotates to predetermined angle  $R$  (for example  $5^\circ$  CA) instead of every fixed number. That is, a timing signal arises every timing in which the crankshaft rotates to a predetermined angle  $R$ , the filter-processed data are integrated during a time period from the timing signal arising to the next timing signal arising.

25        Consequently the number of the integrated value is restrained to a proper number to detect the knocking. "CA" is referred to a rotational angle of the crankshaft (crank angle).

In the case of integration of the filter-processed data every period in which a crankshaft rotates to predetermined angle, another problem arises.

The filter-processed data every sampling timing is grouped every time period of crank angle. The number of the filter-processed data which are integrated in a predetermined angle ( $5^{\circ}$  CA) fluctuates as shown Fig. 14A. Dots in Fig. 14 represent a sampling timing (A/D conversion timing) of the knock sensor signal and a timing of calculation of new filter processed data.

When  $5^{\circ}$  CA indicates  $85 \mu\text{s}$  and a sampling period of the knock sensor signal indicates  $20 \mu\text{s}$  ( $50 \text{ k Hz}$ ), four or five data are integrated are integrated in the  $5^{\circ}$  CA period. The number of the filter-processed data in an integrated value fluctuates.

As shown in Fig. 14B and 14C, the number of filter processed data fluctuates due to an acceleration or deceleration of the engine. "NE" in Fig. 14B and 14C represents a timing signal which traverses every  $5^{\circ}$  CA.

When the number of the filter-processed data in each integrated value fluctuate, the integrated values also fluctuates due to the other condition besides the wave of the knock sensor signal. Thus, the knocking cannot be detected precisely and the feature of the knocking can no be obtained precisely.

#### SUMMARY OF THE INVENTION

An object of the present invention is to reduce the number of the stored data and to detect the knocking precisely.

A knocking detecting device of the present invention has a

signal processing means which converts an analog knock sensor signal into a digital knock sensor signal and filters the digital knock sensor signal digitally every sampling period.

A calculating means of the invention calculates the number  
5 of the A/D conversion for rotating of the crankshaft to a predetermined angle and converts the number into integer  $N$ . Dividing the period for rotating of the crankshaft to the predetermined angle by a sampling period derives the number of the A/D conversion.

An integrating means of the invention integrates the  
10 filter-processed data processed by the signal processing means each  $N$  and a knock detecting means detects the existence of the knocking based on the plural integrated values. The calculating means derives  $N$  before the integrating means starts to process and  $N$  is not updated during the processing period of the integrating means.

15 The integrating means integrates  $N$  pieces of filter-processed values corresponding to the period in which the crankshaft rotates to a predetermined angle. The value of  $N$  is constant during the processing of the integrating means regardless of a start timing of integration and the acceleration or deceleration of the engine.  
20 Therefore, the knocking detection is carried out based on  $N$  pieces of the filter-processed data regardless of a start timing of integration and the acceleration or deceleration of the engine.

The period in which the integrating means performs is defined as following two examples.

25 Example 1: The period is defined as a period from a starting time of processing to a time when the number of integrated value becomes " $M$ ". " $M$ " is a natural number larger than 2.

Example 2: a period from a starting time of processing to a time when the crankshaft rotates to a predetermined angle.

In both of the examples,  $N$  is not fixed number but is established according to the rotational number of the engine. Thus, it is avoided  
5 that the number of stored data for detecting the knocking is increased in the low rotational speed of the engine and the number of stored data is decreased in the high rotational speed of the engine for precise knocking detection.

According to the present invention, the precise knocking  
10 detection is carried out with small number of stored data.

When the knock detecting means detects the knocking according to " $P$ " pieces of integrated value, the knock detecting period is the period in which the crankshaft may rotate to the angle: "the predetermined angle"  $\times$  " $P$ ". During the period, it is determined  
15 whether the knocking exists or not.

#### BRIEF DESCRIPTION OF THE DRAWINGS

Fig. 1 is a block diagram showing an engine control system in a first embodiment of the invention;

20 Fig. 2 is a time diagram showing a processing executed by a microcomputer mounted on the engine control system in the first embodiment;

Fig. 3 is a time diagram showing a processing per one ignition;

Fig. 4 is a flow diagram showing a TDC falling process executed  
25 by the microcomputer in the first embodiment;

Fig. 5 is a flow diagram showing a gate open timing processing executed by the microcomputer in the first embodiment;

Fig. 6 is a flow diagram showing an A/D finishing processing executed by the microcomputer in the first embodiment;

Fig. 7 is a graph showing a relation between a rotation number of an engine (a lateral axis) and a numeral N (a longitudinal axis)  
5 when an engine has six cylinders and an A/D conversion is executed at every 10  $\mu$ s;

Fig. 8 is a flow diagram showing a TDC falling process executed by the microcomputer in a second embodiment of the invention;

Fig. 9 is a chart showing the operation when the rotation number  
10 of an engine is substantially constant in the second embodiment;

Fig. 10 is a chart showing the operation when the rotation number of an engine is rapidly decreased or increased in the second embodiment;

Fig. 11 is a chart showing the operation when the rotation  
15 number of engine is substantially constant in a third embodiment;

Fig. 12 is a chart showing the operation when the rotation number of an engine is rapidly decreased or increased in the third embodiment;

Fig. 13 is a first graph for explaining a conventional system  
20 and the problem thereof;

Fig. 14 is a second graph for explaining a conventional system and the problem thereof.

#### DESCRIPTION OF THE PREFERRED EMBODIMENTS

25 Fig. 1 is a block diagram showing an engine control system  
10 which controls a gasoline engine (an internal combustion engine) having six cylinders.

The engine control system 10 is provided with a multiplexer (MPX) 3, an A/D converter 4, a digital filtering processor 5, a summation processor 6, a knocking determine processor 7, and a CPU 9.

5       The engine control system 10 receives a first analog signal (referred as the first knock sensor signal hereinafter) and a second analog signal (referred as the second knock sensor signal hereinafter). Knock sensors SN1 each of which is respectively disposed on the three cylinders in six cylinders detect the first  
10 knock sensor signal. In this embodiment, the knock sensors SN1 are disposed on a first cylinder # 1, a third cylinder # 3, and a fifth cylinder # 5. Knock sensors SN2 each of which is respectively disposed on other three cylinders detect the second knock sensor signal. In this embodiment, the knock sensor SN2 are disposed on  
15 a second cylinder # 2, a fourth cylinder # 4, and a sixth cylinder # 6.

The multiplexer (MPX) 3 receives one of the first knock sensor signal or second knock signal selectively. The A/D converter 4 converts the signal outputted from the multiplexer 3 during a sampling  
20 period  $t_s$ . In this embodiment, the sampling period is  $10 \mu$  seconds. The digital filtering processor 5 filters the signal converted by the A/D converter 4 digitally. The summation processor 6 adds the data filtered by the digital filtering processor 5, which adds absolute value of the data. The knocking determine processor 7  
25 determines if the knocking arises or not according to the summed value. The CPU 9 controls an ignite timing and an amount of fuel and the like according to the result of the knocking determine



processor 7, the rotational number of the engine, a coolant temperature and other condition of the engine.

The digital filtering processor 5, the summation processor 6 and the knocking determine processor 7 are comprised of microcomputer 8 having CPU, ROM, and RAM.

The engine control system 10 receives a rotational number signal and a cylinder-determining signal. The rotational number signal is for detecting a rotational speed and a rotational position of a crankshaft, which is a pulse signal per a predetermined crankshaft angle. The cylinder determining signal is for determine which cylinder should be ignited, which is a pulse signal per two rotations of the engine 2.

A signal generate circuit 11 generates a TDC signal. As shown in uppermost of Fig. 2, the TDC signal has a falling edge at the timing of each piston top dead center position. The signal generating circuit 11 outputs the TDC signal to the microcomputer 8 and the CPU 9. In this embodiment, since the engine has six cylinders, the TCD signal falls every 120-degree of the crankshaft angle.

Referring to Figs. 2 - 6, the processing of the microcomputer 8 as the digital filtering processor 5, the summation processor 6 and the knocking determine processor 7 is described herein after.

Fig. 2 is a time diagram showing a processing executed by a microcomputer and Fig. 3 is a time diagram showing a processing per one ignition, which shows especially an area surrounded by a broken line in Fig. 2.

The multiplexer 3 selects the first knock sensor signal or the second knock sensor signal and outputs the signal to the A/D

converter 4 at every timing of the falling edge of the TDC signal.  
For example, the knock sensor signal outputted the A/D converter  
4 is changed from the second knock sensor signal to the first knocking  
sensor signal corresponding to the first cylinder # 1 at timing  
5 of TDC of the first cylinder # 1. The knock sensor signal outputted  
the A/D converter 4 is changed from the first knock sensor signal  
to the second knocking sensor signal corresponding to the second  
cylinder # 2 at timing of TDC of the first cylinder # 2.

The knock sensor signal can be changed at the timing of the  
10 gate finishing to the expected knock sensor signal corresponding  
to the next cylinder. The timing of the gate finishing is the timing  
of finishing adding the filtered result data. In such a processing,  
the period from the changing of the knock sensor to the opening  
gate timing is kept enough. That is, the period from the changing  
15 the knock sensor signal inputted to the A/D converter 4 from the  
multiplexer 3 to the opening gate timing is enough. Though there  
is a delay from the changing the knock sensor signal to the getting  
the stable filtered data, the period from the changing of the knock  
sensor signal to the gate opening timing of the next the cylinder  
20 exceeds the delay when the knock sensor signal is changed to the  
signal corresponding to the next cylinder and inputted to the A/D  
converter 4 at the timing of gate finishing before the TDC timing.

To explain the embodiment easily, the first knock sensor signal  
and the second knock sensor signal are both referred as the knock  
25 sensor signal.

In this embodiment, the knocking determining period is from  
the timing of 10-degree CA (timing of ATDC 10-degree CA) to the

timing of 60-degree CA. ATDC means the angle of crankshaft after the timing of TDC.

Fig. 4 is a flow diagram showing a processing just falling of TDC signal. When the microcomputer 8 begins to start the processing, a value of a free run timer is memorized in memory t1 of RAM in step S110. The free run timer is of a count-up timer which counts up by means of the internal clock of the microcomputer 8.

In step S120, a value memorized in a memory t2 of RAM is subtracted from the value memorized in the memory t1 and the value (t1 - t2) is obtained. Next, the TDC period T corresponding to 120-degree CA from the previous TDC timing to present TDC timing is calculated by multiplying the value (t1 - t2) by one count up time tck (one period of the internal clock) as shown in Fig. 2. A value of the free run timer at the previous TDC timing is memorized in the memory t2.

In step 130, a value Z is calculated by a following equation. The period to rotate the crankshaft in a predetermined angle (in this embodiment, the angle is 5-degree) is divided by the sampling period ts (= 10 μs) and the value Z is obtained. Furthermore, the value Z is converted into an integer N by rounding.

$$\begin{aligned} Z &= T \times 0.004167 \\ &= T \times (5^\circ \text{ CA} / 120^\circ \text{ CA} / 10 \mu\text{s}) \quad [\text{equation 1}] \end{aligned}$$

The integer N is a whole number of A/D converting number (sampling number; number of sampling data) corresponding to 5° CA based on the rotational speed of the engine (= TDC period T in this embodiment). Rounding up or rounding down the value Z can derive the integer N.

In step S140, the value of memory t1 is copied to the memory t2. In step S150, the count number  $\alpha$  is calculated based on the following equation with value  $(= t1 - t2)$ , which corresponds to 10° CA. The value of the memory t1 is added to the count number  
5  $\alpha$ , and the added value is memorized as a gate starting time.

$$\alpha = (t1 - t2) \times (10^\circ \text{ CA} / 120^\circ \text{ CA}) \quad [\text{equation 2}]$$

The gate starting time is an estimated time in which the  
10 crankshaft rotates 10° from the present TDC timing based on the rotation speed of engine (= TDC period T in this embodiment). After the process of step S150 is executed, the falling process of the TDC is completed.

Fig. 5 is a flow diagram showing a gate starting timing process  
15 which is executed at the time when the value of the free run timer is in consonance with the gate starting time memorized in step S150.

When the microcomputer 8 starts to execute the gate starting time process, a flag flg is turned on in step 210 and then the value of the memory for summation SUM and counter for summation CNT (referred  
20 as counter CNT herein after) are cleared to zero. In step S230, the value N calculated in step S130 is memorized in a memory n of RAM and the process of the gate starting time process is completed.

When the system can detect the crank angle from the TDC timing based on the rotational signal and the like, ATDC 10° CA is established  
25 instead of the gate starting time, the process shown in Fig. 5 can be executed based on ATDC 10° CA.

Fig. 6 is a flow diagram showing the A/D finishing process

which is executed at every time when the A/D conversion of the knock sensor signal is completed.

When the microcomputer 8 starts to execute the A/D finishing process, a digital filter process is executed and the digital filtered  
5 data is memorized in a memory filt of RAM. The digital filter process is of FIR or IIR digital filter process in which the newest time series A/D converted value is digitally filtered.

In step S320, it is determined whether the flag flg is ON or OFF. When the flag flg is not ON (in S320: NO), the A/D finishing  
10 process is completed. When the flag flg is ON (in S320: YES), the filter-processed data is integrated in step S330.

In step S330, the absolute value of the memory filt is added to the value of the integrate memory SUM and the added value is re-memorized in the integrate memory SUM.

15 In step S340, the value of the memory n is decremented by 1 (-1), and it is determined whether the value of memory n is more than zero in step S350.

When it is determined the value of the memory n is more than zero (in S350: YES), the A/D finishing process is completed. When  
20 the value of the memory n is not more than zero (that is, the value of the memory n is zero), step 360 is executed. In step 360, the present value of the integrate memory SUM is memorized as an integrated value INT (CNT) shown by the counter CNT.

For example, when the value of the counter is zero, the value  
25 of the integrate memory SUM is memorized as a first integrate value INT: zero. When the value of the counter is 1, the value of the integrate memory SUM is memorized as a second integrate value: 1.

In this embodiment, the first value of the counter CNT is zero in step S220, the first integrate value is represented as INT (0).

In step S370, the value of the integrate memory SUM is cleared to zero, and the value N calculated in step S130 is re-memorized  
5 and the value of counter CNT is incremented (+1).

Next, it is determined whether the value of the counter CNT is less than the predetermined value (in this embodiment, the value is 12). When the value of the counter CNT is less than 12, the A/D finishing process is completed.

10 When the value of the counter CNT is not less than 12 in step S390 (that is, the value of the counter CNT reaches 12), the flag flg is turned off in step S400.

In step S410, it is determined whether knocking occurs or not based on the plural integrate values memorized in step S360. In  
15 this embodiment, it is based on from the first integrate value INT (0) through the twelfth integrate value INT (11). The occurrence of the knocking is determined based on the fluctuations in the number of the integrate values, maximum value among the integrate values and the like. Then, the A/D finishing process is completed.

20 In the first embodiment, the TDC interval (interval of the TDC signal falling at every 120° CA) is measured as the rotational speed of the engine in step S111, S120, S140. The value Z is calculated in step S130 and, the value N is derived by rounding the value Z.

When the period from the TDC timing of the first cylinder #  
25 1 to the TDC timing of the second cylinder # 2 is 4000  $\mu$ s as shown in Fig. 2, the value N is 17 which is calculated at the TDC timing of the second cylinder # 2. When the period from the TDC timing

of the second cylinder # 2 to the TDC timing of the third cylinder # 3 is 3900  $\mu$ s, the value  $N$  is 16 which is calculated at the TDC timing of the third cylinder # 3. The timing of a software process shown in lowest portion of Fig. 2 represents the timing in which  
5 the TDC falling process is executed. After a laps of time which corresponds to 10° CA indicated by the counter value  $\alpha$ , the gate starting time process is executed, the flag flg is turned ON (S210), the value of the integrate memory SUM and the counter CNT are cleared to zero (S220), and the value  $N$  at previous TDC timing calculated  
10 in step S130 is memorized in the memory  $n$  of RAM (S230).

After the time when the flag is turned ON, the A/D conversion of the knocking sensor signal by the A/D converter 4 is completed, the digital filter processing is executed in step S310, it is determined YES in step S320, and the filter processed data are  
15 integrated by number of  $N$  according to steps S330 - S390.

In the A/D finishing process of Fig. 6, the filter-processed data are integrated. The number of integrate process is counted by decreasing the value of the memory  $n$  from  $N$  to zero by 1 in the process of step S340 and S350. At every time when the integrate  
20 process is executed  $N$  times (the value of the memory  $n$  is zero), the value of the integrate memory SUM is memorized as the integrate value INT (CNT) in step S360, the value of the counter CNT is counted up by 1 in step S380 so that the number of the integrate value is calculated.

25 When the number of the integrate value is 12 (in step S390: NO), the flag flg is turned off (S400), and it is determined whether the knocking is occurred based on the integrate values INT (0) through

INT (11) which are calculated while the flag is turned ON.

When the TDC timing of the next cylinder comes, the value N is renewed by executing the TDC falling process. The process of Fig. 6 is executed based on the renewed value N.

5 Even when the integrate process is not executed, the digital filter process (S310) is executed, because the digital filter process is supposed to FIR type or IIR type and a past value is necessary besides the newest value.

A predetermined value compared with the value of the counter  
10 CNT in step S390 is established as 12 (twelve integrate values is required) in this embodiment. It is because the period of 60° CA from ATDC 10° CA to ATDC 70° CA is established as a knock determine period. Namely, the value N is a number of sampling (= the number of filter processed data) corresponding to 5° CA, and the period  
15 to acquire a integrate value is a period corresponding to 5° CA. Thus, when the knock determination is executed based on the data corresponding to 60° CA, the twelve integrate values are necessary.

In this embodiment, the A/D converter 4 and the processing  
in step 310 in Fig. 6 correspond to a signal processing means of  
20 the present invention, the processing in step S110 through step S140 in Fig. 4 corresponds to calculating means of the invention, the processing in step S330 through S390 corresponds to an integrating means of the invention, and the processing in step S410 in Fig. 6 corresponds to knock determining means of the invention. The timing  
25 of the flag ON corresponds to a start timing of the integrating means.

The engine control system 10 of the first embodiment can reduce



the number of data to be memorized in order to determine the occurrence of the engine knocking precisely.

Each integrate value  $INT(CNT)$  for the knocking determination is an integrated value of  $N$  filter processed data in a period in which the crankshaft rotates in a predetermined angle (in this embodiment:  $5^\circ$ ). The number of filter processed data is  $N$  and this number is constant notwithstanding the start timing of the integration, acceleration and deceleration of the engine while integrate processing is executed. The value of  $N$  is not fixed value and established by the equation 1 based on the rotational number of engine. Therefore, it prevents the number of the data from increasing at low rotational speed of engine and decreasing at high rotational speed of the engine.

Since the TDC period  $T$  is measured as a rotational speed of the engine and TDC period  $T$  is multiplied by the constant number ( $= 0.004167$ ) to get the value  $Z$ , the rotational speed of the engine is not calculated in the [rpm] dimension and load of processing to calculate the value  $N$  is decreased. The TDC period  $T$  is an interval between adjacent falling edges of the TDC signal (standard signal in this invention) which arises every  $120^\circ$  of crankshaft (a constant angle in this invention). The period corresponding to  $5^\circ$  CA is divided by the sampling period to get the value  $Z$ . In stead of the TDC period  $T$ , the period between pulse signals arise at every  $30^\circ$  of the crankshaft can be measured to obtain the value  $Z$  and value  $N$ .

In step S130 shown in Fig. 4 as the calculating means, it is determined whether the TDC period  $T$  calculated in step 120 is between a first predetermined value  $T1$  and a second predetermined value

T2. (Case 1): In case of  $T \geq T1$ , the value  $N$  is fixed to value  $N1$ .  
(Case 2): In case of  $T \geq T2$ , the value  $N$  is fixed to value  $N2$  ( $< N1$ )

In the case 1, it is avoided that the value  $N$  exceeds a  
5 predetermined value in low speed of engine and the integrated value  
such as the integrate memory SUM and value INT overflows. When the  
value  $N$  is fixed to the first value  $N1$ , the number of the integrate  
value INT is increased so that the substantial knocking determination  
period is not reduced.

10 In the case 2, accuracy of the integrate value is kept high  
even in the high rotation speed of the engine. When the value  $N$   
becomes small according to the engine speed up, the integrate value  
may be fluctuated and the detecting accuracy of the knocking may  
be decreased. Since the value  $N$  is fixed to the value  $N2$  at high  
15 engine speed, the deterioration of the detecting accuracy is avoided.

Fig. 7 is a graph showing the relationship between the RPM  
of the engine (a lateral axis) and value  $N$  (a longitudinal axis),  
wherein the engine has 6 cylinders, and an A/D conversion is processed  
at every 10  $\mu$ s. When the speed of engine is 1000 rpm and the TDC  
20 period  $T$  is 20 ms, the value  $N$  is 83. When the speed of the engine  
is 327 rpm and the TDC period  $T$  is 61.2 ms, the value  $N$  is 255.  
When the speed of the engine is 8333 rpm and the TDC period  $T$  is  
2.4 ms, the value  $N$  is 83. Therefore, when the speed of engine is  
less than 327 rpm, the value  $N$  is fixed to 255, and when the speed  
25 of the engine is larger than 8333 rpm, the value  $N$  is fixed to 10.

The processing of the value  $N$  can be applied to the other  
embodiment and modifications described below.

In the first embodiment, rounding the value  $Z$  in step S130 shown in Fig. 4 derives the value  $N$ . When the value  $N$  is derived by rounding off the number of decimal places i.e. the number of decimal places of the value  $Z$  is less than 0.5, the number of the  
5 integrate value INT can be increased. For example, in step S390 shown in Fig. 6, the set value 12 can be changed into 13.

When the value  $N$  is derived by rounding off and the number of integrate value is not increased, the period from the timing of the flag ON to the timing of finishing the calculation twelve  
10 integrated values are shorter than the actual period corresponding to  $60^\circ$  CA which is knock determining period. By increasing the number of integrate value from 12 to 13, it is prevented that the actual knock determining period is shorter than the designed knock determining period. In other words, the integrate value  
15 corresponding to  $60^\circ$  CA can be obtained, which is necessary to determine the knocking precisely.

A second embodiment of the invention is described herein after.

The differences between the first embodiment and the second embodiment are following two points.

20 (First difference 1-1)

In step S390, it is determined whether the value of the counter CNT is larger than 17. Therefore, the integrate values INT are calculated from INT (0) through INT (16).

Step S140 is deleted. In the A/D converting process, the  
25 knocking determination is not carried out.

(Second difference 1-2)

The TDC falling process shown in Fig. 8 is carried out instead

of the TDC falling process shown in Fig. 4.

In the TDC falling process shown in Fig. 8, step S150 is modified and step S160 and S170 are added after step S150. The other processes are same as the process shown in Fig. 4.

5           In step S150, according to the subtracted value ( $= t_1 - t_2$ ), a value of the free run timer is calculated. The value of the free run timer as a count value  $\alpha$  corresponds to the  $10^\circ$  CA. The count value  $\alpha$  is added to the value memorized in the memory  $t_1$  to get a new value which is memorized as a gate starting time. In the second  
10   embodiment, the gate starting time represents an estimated time after  $5^\circ$  CA from the TDC timing.

When the rotational speed of the engine is constant, and when there is no difference between the previous TDC timing  $T$  and the present TDC timing  $T$ , seventeen integrate values are calculated  
15   during knock determining period corresponding to  $85^\circ$  CA: from ATDC  $5^\circ$  CA to  $90^\circ$  CA. The integration of  $N$  filter processed data (corresponding to  $5^\circ$  CA) is executed during the ATDC  $5^\circ$  CA to ATDC  $90^\circ$  CA. In Fig. 9 and Figs. 10 - 12, each of the integrate value is illustrated at the previous position compared to Fig. 3.

20           In step S160, a changing rate  $W (= (T_b / T_a) - 1)$  is calculated. The changing rate  $W$  represents the increased amount of the rotational speed of the engine. According to the TDC period  $T$  calculated in the previous step S120 (corresponding to the rotational speed  $V_b$ , hereinafter denoted as  $T_b$ ) and the TDC period  $T$  calculated in present  
25   step S120 (corresponding to the rotational speed  $V_a$ , hereinafter denoted as  $T_a$ ), the rotational speed of the engine  $V_a$  is obtained. The rotational speed of the engine  $V_b$  is obtained from the second

previous TDC timing to the previous TDC timing. The changing rate  $W$  represents the increased amount from the rotational speed  $V_a$  to the rotational speed  $V_b$ .

In step S160, the series number of the integrate value for  
5 the knocking determination is selected from the 17 integrate values according to the changing rate  $W$ .

(Case 1-2-2)

When the changing rate  $W$  is between  $-G$  ( $G$  is positive number, for example 0.1) and  $+G$ , i.e. the increased amount or decreased  
10 amount of the rotational speed of the engine is within  $100 \times G$  percent, and when the rotational speed of the engine is constant, the second through thirteenth integrate value is selected as shown in Fig. 9. In this case, the integrate value corresponding to the ATDC 10  
° CA through ATDC 70° CA is selected.

15 (Case 1-2-2)

When the changing rate  $W$  is less than  $-G$ , i.e. the rotational speed of the engine is decreased more than  $100 \times G$  percent, the second integrate value through the sixteen integrate value are selected as shown in middle part of Fig. 10. In this case, the integrate  
20 value corresponding to the ATDC 10° CA through ATDC 70° CA after deceleration is selected.

(Case 1-2-3)

When the changing rate  $W$  is more than  $+G$ , i.e. the rotational speed of the engine is increased more than  $100 \times G$  percent, the first  
25 integrate value through the eleventh integrate value are selected as shown in middle part of Fig. 10. In this case, the integrate value corresponding to the ATDC 10° CA through ATDC 70° CA after

acceleration is selected.

In step S170, the knocking determination is processed based on the integrate values selected in step S160.

It is unnecessary to derive seventeen integrate value. It is  
5 enough to derive sixteen integrate value: from the first integrate  
value to the sixteenth value. When the rotational speed of the engine  
is constant, twelve integrate values are used for knocking  
determination. When the rotational speed of the engine is decreased,  
fifteen integrate values are used for knocking determination. When  
10 the rotational speed of the engine is increased, eleven integrate  
values are used for knocking determination. These number of the  
integrate values does not affect the recognition of the knocking  
wave. In the second embodiment, the process executed in step S160  
corresponds to a selecting an integrate value means.

15 In the second embodiment, since the plural integrate values  
are selected according to the engine condition, even in acceleration  
or deceleration of the engine, it is possible to determine the knocking  
by means of the integrate value corresponding to the ATDC 10° CA  
through ATDC 70° CA as shown in Fig. 10.

20 When the knocking determination is carried out by recognizing  
the shape of the knock sensor signal wave, latter integrate values  
(for example, tenth integrate value henceforth) do not affect the  
knocking determination, thus following modification can be  
established.

25 In step S160, only when the engine is accelerated in the case  
1-2-3, the first integrate value through the eleventh integrate  
value are selected for the knocking determination. In other case,

the second integrate value through the twelfth integrate values can be selected.

As described above, since the constant number of the integrate value is selected, the knocking determination process in step S170  
5 can be stable. In this case, twelve integrate values are enough to determine the knocking.

When the integrate processing of the knock sensor signal during the knocking determination corresponding to the ATDC 10° CA - 70° CA is needed, the integration corresponding to 5° CA is configured  
10 so that the number of the integrate value become seventeen since the estimated timing of the ATDC 5° CA. After detecting the timing of the ATDC 70° CA which is the timing of gate finishing, the integrate processing corresponding to 5° CA can be finished even if the number of integrate value has not reached seventeen.

15 When the engine is accelerated, the first through the last integrate values are used. In other case, the second integrate value through the last integrate value can be used. By comparing the actual timing of the ATDC 70° CA with the estimated timing of the ATDC 70° CA, it is acquired that the engine is accelerated or decelerated.

20 In this configuration, the actual timing of the ATDC 70° CA is obtained not withstanding the speed of the engine. The way of stopping the integrate processing at the timing of the ATDC 70° CA is to stop the processing after all integration is accomplished and to stop the processing immediately with utilizing the previous  
25 value.

A third embodiment of the invention is described below.

The engine control system of the third embodiment is different

from the second embodiment at following three points (2-1) - (2-3).

(First difference 2-1)

In step S130 shown in Fig. 8, according to the TDC period  $T$ , the value  $Z$  is obtained by dividing the period which is required  
5 for the crankshaft to rotate a certain angle less than  $5^\circ$  (in the first embodiment,  $1^\circ$ ) by the sampling period  $t_s$  ( $= 10 \mu s$ ), and then the value  $N$  is calculated by rounding the value  $Z$ .

The value  $Z$  is calculated by the following equation.

$$Z = T \times 0.000833$$

10 
$$= T \times (1^\circ \text{ CA} / 120^\circ \text{ CA} / 10 \mu s) \quad (\text{equation 3})$$

In the third embodiment, the value  $N$  is set to  $1 / m$  ( $m$  is a natural number 2 or more. In this embodiment,  $m = 5$ ), the filter processed data is added one by one which corresponds  $1^\circ$  CA. The integration of the filter processed data  $N$  is referred as an  
15 integration of  $1^\circ$  CA.

(Second difference 2-2)

The preset value compared to the value of counter CNT in step S390 shown in Fig. 6 is determined to the value ( $= 85$ ) which is derived by multiplying the value ( $= 17$ ) and  $m$  ( $= 5$ ).

20 The integration of  $1^\circ$  CA is derived 85 times from INT (0) to INT (84) in the third embodiment. When the rotational speed of the engine is constant, and when the period from the pervious TDC timing to the present TDC timing is equal to the period from the present TDC timing to the next TDC timing, the integration of  $1^\circ$  CA is derived  
25 85 times in the range from the ATDC  $5^\circ$  CA to the ATDC  $90^\circ$  CA, the range being more broad than the knock determining period (ATDC  $10^\circ$  CA to ATDC  $90^\circ$  CA).



(Third difference 2-3)

In step S160 shown in Fig. 8, the changing rate  $W (= (T_b / T_a) - 1)$  is calculated as same as the second embodiment, however, the process selecting the integrate value is skipped and step S170  
5 is executed.

In step S170, based on the changing rate  $W$ , the knocking determination is processed as following steps.

(Step 1)

When the changing rate  $W$  is more than  $-G$  ( $G$  is a positive  
10 number: for example, 0.1) and less than  $+G$ , i.e. acceleration and deceleration is within the rang of  $100 \times G$  percent and the rotational speed of the engine is substantially constant, each integrate value corresponding to  $1^\circ$  CA is integrated at every " $m$ " value from the first value to get the " $b$ " integrated values. In this embodiment,  
15 " $m$ " is 5 and " $b$ " is 12. Then, it is determined whether the knocking arises or not. In this case, the knocking determination is carried out on the basis of the integrated value corresponding to predetermined period (the ATDC  $10^\circ$  CA to the ATDC  $70^\circ$  CA).

(Step 2)

20 When the changing rate  $W$  is less than  $-G$ , i.e. deceleration of engine is more than  $100 \times G$  percent, the knocking determination is executed according to the twelve integrated values as shown in middle portion in Fig. 12. The twelve integrated values are derived by integrating each integration of  $1^\circ$  CA every certain number which  
25 is lager than " $m$ ", for example, 6.

The knocking determination is processed based on the value corresponding to the ATDC  $10^\circ$  CA to the ATDC  $70^\circ$  CA after deceleration.

(Step 3)

When the changing rate  $W$  is larger than  $+G$ , i.e. acceleration of engine is more than  $100 \times G$  percent, the knocking determination is executed according to the twelve integrated values as shown in lower portion in Fig. 12. The twelve integrated values are derived by integrating each integration of  $1^\circ$  CA at every certain number which is less than " $m$ ", for example, 4.

The knocking determination is processed based on the value corresponding to the ATDC  $10^\circ$  CA to the ATDC  $70^\circ$  CA after acceleration.

10 In the third embodiment, since the plural integrate values are selected according to the engine condition it is possible to determine the knocking by means of the integrate value corresponding to the ATDC  $10^\circ$  CA through ATDC  $70^\circ$  CA as shown in Fig. 12, even in acceleration or deceleration of the engine.

15 The embodiments described above can be modified as follows.

The knocking determining process can be executed by plural digital filter process and integral process. The plural digital filter processes are executed to the A/D converted knock sensor signal, the digital filter processed data are integrated, and then  
20 it is determined whether the knocking arises or not. In this case, the number of the integrated value is changeable at every filtering process. When the number of the integrated value is constant, the counter CNT and the memory  $n$  can be common in each processing.

A vibrating sensor, an ion current sensor or a cylinder pressure  
25 sensor can be used as the knock sensor.